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**UTILITY  
PATENT APPLICATION  
TRANSMITTAL**

(Only for new nonprovisional applications under 37 CFR 1.53(b))

Attorney Docket No.

TI-20390

First Inventor

Sami Kiriaki

Title

ANALOG MULTIPLEXERS WITH CMOS CONTROL SIGNALS

Express Mail Label No.

EL552916238US

**APPLICATION ELEMENTS**

See MPEP chapter 600 concerning utility patent application contents

**ADDRESS TO:**Assistant Commissioner for Patents  
Box Patent Application  
Washington, DC 20231

1. ☒ Fee Transmittal Form (e.g., PTO/SB/17)  
(Submit an original, and a duplicate for fee processing)
2. ☐ Applicant claims small entity status.  
See 37 CFR 1.27.
3. ☒ Specification (preferred arrangement set forth below) [ Total Pages **22** ]
- Descriptive title of the Invention
  - Cross Reference to Related Applications
  - Statement Regarding Fed sponsored R & D
  - Reference to sequence listing, a table, or a computer program listing appendix
  - Background of the Invention
  - Brief Summary of the Invention
  - Brief Description of the Drawings (if filed)
  - Detailed Description
  - Claim(s)
  - Abstract of the Disclosure
4. ☒ Drawing(s) (35 U.S.C. 113) [ Total Sheets **6** ]
5. Oath or Declaration [ Total Pages ]
- a. ☒ Newly Executed (original or copy)
- b. ☐ Copy from a prior application (37 CFR 1.63(d))  
(for continuation/divisional with Box 18 completed)
- i. ☐ **DELETION OF INVENTOR(S)**  
Signed statement attached deleting inventor(s)  
named in the prior application, see 37 CFR  
1.63(d)(2) and 1.33(b)
6. ☐ Application Data Sheet. See 37 CFR 1.76

7. ☐ CD-ROM or CD-R in duplicate, large table or  
Computer Program (Appendix)
8. Nucleotide and/or Amino Acid Sequence Submission  
(if applicable, all necessary)
- a. ☐ Computer Readable Form (CRF)
- b. Specification Sequence Listing on:
- i. ☐ CD-ROM or CD-R (2 copies); or
- ii. ☐ paper
- c. ☐ Statements verifying identity of above copies

**ACCOMPANYING APPLICATION PARTS**

9. ☒ Assignment Papers (cover sheet & documents(s))
10. ☐ 37 CFR 3.73(b) Statement (when there is an assignee) ☐ Power of Attorney
11. ☐ English Translation Document (if applicable)
12. ☐ Information Disclosure Statement (IDS)/PTO-1449 ☐ Copies of IDS Citations
13. ☒ Preliminary Amendment
14. ☒ Return Receipt Postcard (MPEP 503)  
(Should be specifically itemized)
15. ☐ Certified Copy of Priority Document(s)  
(if foreign priority is claimed)
16. ☐ Request and Certification under 35 U.S.C. 122  
(b)(2)(B)(i). Applicant must attach form PTO/SB/35  
or its equivalent
17. ☐ Other.

18. If a CONTINUING APPLICATION, check appropriate box, and supply the requisite information below and in a preliminary amendment, or in an Application Data Sheet under 37 CFR 1.76:

☐ Continuation☐ Divisional☐ Continuation-in-part (CIP)

of prior application No. \_\_\_\_\_ / \_\_\_\_\_

Prior application information

Examiner \_\_\_\_\_

Group / Art Unit \_\_\_\_\_

For CONTINUATION OR DIVISIONAL APPS only: The entire disclosure of the prior application, from which an oath or declaration is supplied under Box 5b, is considered a part of the disclosure of the accompanying continuation or divisional application and is hereby incorporated by reference. The incorporation can only be relied upon when a portion has been inadvertently omitted from the submitted application parts.

**19. CORRESPONDENCE ADDRESS**

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**FEE TRANSMITTAL**

Patent fees are subject to annual revision on October 1.

These are the fees effective November 10, 1998

Small Entity payments must be supported by a small entity statement, otherwise large entity fees must be paid. See Forms PTO/SB/09-12**Complete If Known**

Application Number	
Filing Date	11/13/00
First Named Inventor	Sami Kiriaki
Examiner Name	TBD
Group / Art Unit	TBD
Attorney Docket No.	TI-20390

TOTAL AMOUNT OF PAYMENT

(\$)**710****METHOD OF PAYMENT**

1. ☒ The Commissioner is hereby authorized to charge to the following Deposit Account,

Deposit Account Number

20-0668

Deposit Account Name

Texas Instruments Incorporated

- ☒ Charge any additional fee required or credit any overpayment ☐ Charge all indicated fees and any additional fee required or credit any overpayment

2. ☐ **Payment Enclosed:**

☐ Check ☐ Money Order ☐ Other

**FEE CALCULATION****1. BASIC FILING FEE**

Large Fee Code	Entity Fee (\$)	Small Fee Code	Entity Fee (\$)	Fee Description	Fee Paid
101	710	201	380	Utility filing fee	710
106	310	206	155	Design filing fee	\$
107	480	207	240	Plant filing fee	\$
108	760	208	380	Reissue filing fee	\$
114	150	214	75	Provisional filing fee	\$

SUBTOTAL (1) (\$)**710****2. EXTRA CLAIM FEES**

Total Claims	Independent Claims	Multiple Dependent	Extra Claims	Fee from below	Fee Paid
10	3		-20** = 0	18	0
			-3** = 0	78	0
				260	0

\*or number previously paid, if greater; For Reissue, see below

Large Fee Code	Entity Fee (\$)	Small Fee Code	Entity Fee (\$)	Fee Description
103	18	203	9	Claims in excess of 20
102	78	202	39	Independent Claims in excess of 3
104	260	204	130	Multiple dependent claims in excess of 3
109	78	209	39	**Reissue independent claims over original patent
110	18	210	9	**Reissue claims in excess of 20 and over original patent

SUBTOTAL (2) (\$)**0****FEE CALCULATION (continued)****3. ADDITIONAL FEES**

Large Fee Code	Entity Fee (\$)	Small Fee Code	Entity Fee (\$)	Fee Description	Fee Paid
105	130	205	65	Surcharge - late filing fee	
127	50	227	25	Surcharge - late provisional filing fee or cover sheet	
139	130	139	130	Non-English specification	
147	2,520	147	2,520	For filing a request for reexamination	
112	920*	112	920*	Requesting publication of SIR prior to Examiner action	
113	1,840*	113	1,840*	Requesting publication of SIR after Examiner action	
115	110	215	55	Extension for reply within first month	
116	380	216	190	Extension of time within second month	
117	870	217	435	Extension of time within third month	
118	1,360	218	680	Extension of time within fourth month	
128	1,850	228	925	Extension of time within fifth month	
119	300	219	150	Notice of Appeal	
120	300	220	150	Filing a brief in support of an appeal	
121	260	221	130	Request for oral hearing	
138	1,510	138	1,510	Petition to institute a public use proceeding	
140	110	240	55	Petition to revive - unavoidable	
141	1,210	241	605	Petition to revive - unintentional	
142	1,210	242	605	Utility issue fee (or reissue)	
143	430	243	215	Design issue fee	
144	580	244	290	Plant issue fee	
122	130	122	130	Petitions to the Commissioner	
123	50	123	50	Petitions related to provisional applications	
126	240	126	240	Submission of Information Disclosure Stmt	
581	40	581	40	Recording each patent assignment per properly (time number of properties)	
146	790	246	395	Filing a submission after final rejection (37 CFR 1.129(a))	
149	790	249	395	For each additional invention to be examined (37 CFR 1.129(b))	

Other fee (specify)

Other fee (specify)

\*Reduced by Basic Filing Fee Paid

SUBTOTAL (3)

**SUBMITTED BY**

Typed or Printed Name

Carlton H. Hoel

Signature

Date

11/13/00

**Complete (if applicable)**

Reg Number

29,934

Deposit Account User ID

**IN THE UNITED STATES PATENT AND TRADEMARK OFFICE**

In re the Application of:

Sami Kiriaki  
Serial No.: TBD  
Filed: 11/13/00  
For: **ANALOG MULTIPLEXERS WITH CMOS CONTROL SIGNALS**

Docket No.: TI-20390  
Examiner: TBD  
Art Unit: TBD

**PRELIMINARY AMENDMENT**

Honorable Assistant Commissioner

of Patents

Washington, D.C. 20231

"EXPRESS MAILING" Mailing Label No. EL552916238US

Date of Deposit: November 13, 2000.

Please amend the above referenced Application as follows:

**In the Specification:**

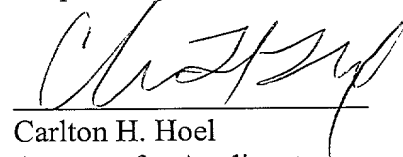
Page 1, before line 1, insert --This application claims priority under 35 USC §119(e)(1) of provisional application number 60/164,996 filed 11/12/99.--

**REMARKS**

Entry of the foregoing amendment prior to examinations is respectfully requested.

If the Examiner has any questions or other correspondence regarding this application, Applicant requests that the Examiner contact Applicant's attorney at the below listed telephone number and address.

Respectfully submitted,



Carlton H. Hoel  
Attorney for Applicant  
Reg. No. 29,934

Texas Instruments Incorporated  
P.O. Box 655474, MS 3999  
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# ANALOG MULTIPLEXERS WITH CMOS CONTROL SIGNALS

## BACKGROUND OF THE INVENTION

### 5 1. Field of the Invention

This invention relates to fast analog multiplexers with CMOS control signals and particularly to the elimination of cross-signal feed-through in these high  
10 speed circuits.

### 2. Brief Description of the Known Art

Figure 1a shows the simplest type of conventional  
15 analog multiplexer 1 (MUX) built with CMOS switches. Here, the MUX switches are comprised of n-channel MOS transistors 2-4 and p-channel MOS transistors 5-7, connected in parallel to form a CMOS switch. This shows a n-to-1 MUX with input signals sig 1, sig 2, ... sig n  
20 connected to MUX switches 2/5, 3/6, and 4/7, respectively. The output of the MUX switches are connected together and become the output of the MUX circuit. The MUX switches are controlled by placing complementary control voltages on the transistor gates, as shown. The logic circuit for  
25 generating these control signals is comprised of two inverters 8 and 9, as shown in Figure 1b. As example, for operation between 0 volts and positive power supply  $V_{DD}$ ,

the switches are turned "ON" and "OFF" at their gates, as follows:

5		n-Channel Transistor	p-Channel Transistor
	ON	$V_{DD}$	0V
	OFF	0V	$V_{DD}$

where  $V_{DD}$  is the positive power supply (e.g. +5V) and 0V is ground.

10 By having the n-channel and p-channel transistors connected in parallel, the circuit can handle signals from 0 to  $V_{DD}$  volts, as illustrated in Figure 1c. In this figure,  $V_{sig}$  is plotted on the abscissa and the ON resistance of the switches,  $R_{SW}$ , is plotted on the ordinate. The transistors tend to be "ON" over the following voltage ranges, respectively:

$$\text{n-channel:} \quad 0 \rightarrow [V_{DD} - V_{th(n-ch)}]$$

$$\text{p-channel} \quad V_{DD} \rightarrow [0 + V_{th(p-ch)}]$$

This means that for small signals the n-channel transistor is primarily used and for large signals the p-channel transistor is primarily used. Also, the ON resistance,  $R_{SW}$ , tends to be optimal (lowest) in mid-signal range where both parallel transistors are ON.

The primary problem with analog multiplexers of this type, used to select one of several input signals, is that they often have undesirable signal feed-through where an  
5 attenuated level of an unselected signal appears as part of the output signal. This feed-through is due primarily to the parasitic capacitances,  $C_{gd}$  and  $C_{gs}$ , associated with the CMOS transistors used to implement the switches. As a result, this undesirable feed-through causes a  
10 degradation at the output of both the signal-to-noise ratio (SNR) and the signal-to-distortion ratio (SDR) for the selected signal at the output of the MUX.

Thus, there is a need for an improved high speed MUX  
15 which eliminates the cross-signal feed-through problems of the prior art. The invention and embodiment disclosed herein address this need.

For reference, U. S. Patents 5,744,995 discusses  
20 multi-input multiplexers and U. S. patent 5,598,114 discusses high-speed multiplexers.

## SUMMARY OF THE INVENTION

This invention addresses the shortcomings of prior art  
5 analog multiplexers, depending on the application, to  
provide low-distortion, high-speed solutions. The objective  
is to provide high-speed multiplexers which eliminates  
cross-signal feed-through at the circuit's output. These  
designs take into account such parameters as input signal  
10 level, signal bandwidth, common mode operation, parasitic  
capacitance, and transistor layout.

The circuits of this invention use N-MOS/P-MOS  
transistor pairs for signal switches and additional N\_MOS  
15 transistors to effectively shunt the unselected signal paths  
to circuit ground, thereby considerably reducing the amount  
of undesired signal presence at the circuit's output.

Two embodiments of the invention address the signal  
20 feed-through issue with CMOS circuitry; one for limited  
bandwidth applications and one for small signal  
applications. For small signal applications, the P-MOS  
transistors in the signal switches of these circuits are  
eliminated, leaving only the N-MOS transistors, to provide  
25 improved bandwidth and lower signal feed-through. All of  
the techniques of this invention can be applied to both  
single-ended and/or differential configurations.

Also, the layout of CMOS transistors with reduced parasitic capacitance, used in the implementation of the circuits of this invention, are included in the discussion.



## BRIEF DISCRIPTION OF THE DRAWINGS

The included drawings are as follows:

5 Figure 1a shows a schematic for a simple analog MUX with CMOS control. (prior art)

Figure 1b shows a typical schematic of the control logic for an analog MUX. (prior art)

10 Figure 1c illustrates the switching characteristics for the n-channel and p-channel transistors commonly used in the analog MUX circuitry. (prior art)

Figure 2 shows the schematic for a CMOS MUX of this invention with improved input signal feed-through.

15 Figure 3 shows the schematic for a small-signal, high-bandwidth, single-ended CMOS MUX of this invention with improved input signal feed-through.

Figure 4 shows the schematic for a small-signal, high-bandwidth, differential CMOS MUX of this invention with improved input signal feed-through.

20 Figures 5a and 5b illustrate the layout of the CMOS transistors of this invention with reduced parasitic output capacitance for use in analog multiplexer applications.

25

## DESCRIPTION OF THE PREFERRED EMBODIMENT

Figure 2 shows one embodiment for a CMOS MUX 10 with improved signal feed-through characteristics. The circuit is comprised of an input pair of n-channel 11-13/p-channel 17-19 MOS transistor switches and an output pair of n-channel 14-16/p-channel 20-22 MOS transistor switches connected in series at each input and additional n-channel MOS transistors 23-25 connected, as pull-down devices, between each series transistor pair and circuit ground. Input signals Sig<sub>1</sub>, Sig<sub>2</sub>, and Sig<sub>n</sub> are connected to transistors pairs 11/17, 12/18, and 13/19, respectively. The outputs of MOS transistor pairs 14/20, 15/21, and 16/22 are tied together to form a low feed-through output signal. Each of the series n-channel MOS transistors 11-16 are driven at the gate by logic control signals,  $S_n$ , while the series p-channel MOS transistors 17-22 and n-channel MOS pull-down transistors 23-25 are driven by logic control signals,  $\overline{S_n}$ , which are complementary to the above 11-16. As a result, the overall effect of this circuit arrangement is as though there were individual switches, with low signal feed-through, for each input signal.

In operation, when one signal is selected, all other signals are shunted to ground by their associated pull-

down transistors, such that feed-through from the  
unselected signals is eliminated at the output. For  
example, if Sig<sub>2</sub> is selected, then MOS pull-down  
transistor switch 24 is OFF, allowing the Sig<sub>2</sub> signal to  
5 pass through to the output while MOS pull-down  
transistors 23 and 25 are ON, shunting any feed-through  
from signals Sig<sub>1</sub> and Sig<sub>n</sub> to ground and preventing any  
feed-through of these unselected signals at the output.  
Either the n-channel or p-channel MOS transistor can be  
10 selected as the ON switch, depending on the level of the  
input signal.

This circuit is limited to rather low bandwidth  
applications due to the total RC time constant associated  
15 with each switch. For example, switch SW<sub>1X</sub> 11/17 has an  
ON resistance of R<sub>1X</sub> and a total parasitic capacitance  
C<sub>1X</sub> at node N<sub>1</sub> and switch SW<sub>1Y</sub> 14/20 has an ON resistance  
of R<sub>1Y</sub> and a total parasitic capacitance C<sub>1Y</sub> at the  
output node. Therefore, the total RC time constant for  
20 Sig<sub>1</sub> is given as:

$$R_{1X} \cdot C_{1X} + R_{1Y} \cdot C_{1Y}$$

For a given switch control level and common mode signal,  
25 the switch ON resistance can be reduced by increasing the  
widths of both the N-MOS and P-MOS transistors. However,

this reduction in ON resistance is typically accompanied by an increase in the drain-to-bulk and source-to-bulk parasitic capacitance. But, an optimum design can be found for a limited number of signals that are joined  
5 together at the MUX output for a given application.

In a second embodiment of the circuit 26, for the case of small signal applications where the input signal is a small fraction of the MUX supply voltage, the switch  
10 bandwidth can be improved by modifying the circuit as shown in Figure 3. Since the mobility of electrons is approximately three times greater than that for holes, the problem with the ON switch resistance discussed above is magnified by the fact that P-MOS device sizes must be  
15 made three times or more the size of the N-MOS devices to overcome this mobility difference. However, this larger size for the P-MOS devices results in larger parasitic capacitance which in turn increases the RC time constant and reduces the switch bandwidth. In this circuit, used  
20 primarily for small signal applications, the p-channel MOS transistors 17-22 (in Figure 2) are eliminated completely so that the series switches consists of only n-channel MOS transistor switches 11-16 and the n-channel MOS shunting transistors 23-25. Otherwise, the circuit  
25 configuration is the same as in Figure 2. In this circuit the parasitic capacitance is reduced by as much

as 50%, assuming the signal common mode level is closer to the MUX ground and the peak-to-peak signal level is a small fraction of the MUX supply voltage. This configuration is applicable as long as the signal voltage level ( $V_{cm} + V_{sig}$ ) produces enough  $V_{gs}$  for the series N-MOS transistors 11-16 to maintain a low ON resistance.

The circuits discussed above are shown for single-ended signal applications. However, all these circuits can be implemented for fully differential operation, as illustrated in Figure 4 for the small signal circuit discussed above. This circuit is essentially comprised of two of the single-ended circuits 26 coupled so as to provide for differential inputs signals  $Sig_1+/Sig_1-$ ,  $Sig_2+/Sig_2-$ ,  $Sig_n+/Sig_n-$  and a differential output signal  $Sig_o+/Sig_o-$ .

The parasitic capacitance in these high-speed MUX circuits can be further reduced by using an even number of "fingers" in the circuit layout of the series output transistors SW1y 14, SW2y 15, SWny 16, as illustrated in Figures 5a and 5b. Figure 5a shows a typical layout for a single finger N-MOS transistor which is comprised of a source 27, a gate 28, and a drain 29. In this case the source and drain capacitances are equal and given as the

area of the gate width (w) and drain/source length (x),  
as follows:

$$C_S = C_D = x \cdot w$$

5

On the other hand, for the lower capacitance layout of  
this invention, shown in Figure 5b, the gate and source  
are split into two parts, or "fingers", so that the  
transistor is comprised of sources 30 and 31, gates 32  
and 33, and drain 34, and the capacitance-to-area  
relationship becomes:

10

$$C_D = x \cdot \frac{w}{2}, \quad \text{and}$$

$$C_S = 2 \cdot x \cdot \frac{w}{2} = x \cdot w, \quad \text{so that}$$

15

$$C_S = 2 \cdot C_D.$$

This means that a two "finger" device has a drain-to-bulk  
parasitic capacitance,  $C_{dB}$ , that is one-half the source  
capacitance,  $C_S$ , and as a result the total output  
capacitance is reduced by at least 50%. This layout can  
be used to obtain a significant boost, where the  
bandwidth of the MUX circuit is at least doubled.

20

While the invention has been described in the context  
of two preferred embodiments, it will appear to those  
skilled in the art that the present invention may be

25

modified in numerous ways and may assume many embodiments  
other than that specifically set out and described above.  
Accordingly, it is intended by the appended claims to  
cover all modifications of the invention which fall within  
5 the true spirit and scope of the invention.

I CLAIM:

1. A CMOS analog multiplexer circuit comprising:  
multiple series n-channel/p-channel MOS transistor  
5 input switches;  
multiple series n-channel/p-channel MOS transistor  
output switches;  
multiple pull-down n-channel MOS transistor switches;  
wherein  
10 said circuit prevents cross-signal feed-through from  
unselected inputs by shunting said feed-through to  
circuit ground.
2. The CMOS analog multiplexer circuit of Claim 1  
15 further comprising:  
the input of a first series input n-channel/p-  
channel MOS transistor switch coupled to the first  
circuit input signal;  
the input of a second series input n-channel/p-  
20 channel MOS transistor switch coupled to the second  
circuit input signal;  
the input of a nth series input n-channel/p-channel  
MOS transistor switch coupled to the nth circuit  
input signal;  
25 the output of said first series input n-channel/p-  
channel MOS transistor switch coupled to the input



of a first output n-channel/p-channel MOS transistor switch and to the drain of a first n-channel MOS pull-down transistor;

5 the output of said second series input n-channel/p-channel MOS transistor switch coupled to the input of a second output n-channel/p-channel MOS transistor switch and to the drain of a second n-channel MOS pull-down transistor;

10 the output of said nth series input n-channel/p-channel MOS transistor switch coupled to the input of a nth output n-channel/p-channel MOS transistor switch and to the drain of a nth n-channel MOS pull-down transistor;

15 the outputs of said first, second, and nth output n-channel/p-channel MOS transistor switches coupled together and to circuit output;

the sources of said first, second, and nth n-channel MOS pull-down transistors coupled to circuit ground;

20 the gates of all said series input and output n-channel MOS transistor switches coupled to a logic control signal;

the gates of all said series input and output p-channel MOS transistor switches and all n-channel MOS pull-down transistors coupled to a logic control

signal which is the complement of said input/output transistor logic control signal.

3. The CMOS analog multiplexer circuit of Claim 2,  
5 further comprising MOS transistor layouts which have an even number of circuit fingers: wherein said MOS transistor's parasitic capacitance is reduced by more than 50%; and said MOS transistor's bandwidth is at least doubled.
- 10 4. The CMOS transistor layout of Claim 3 further comprising:  
a CMOS transistor whose architecture has the source and gate split in half such that it consists of the  
15 sequence: one-half source, one-half gate, drain, one-half gate, and one-half source;  
said CMOS transistor whose drain capacitance is half that of a conventional CMOS transistor;  
said CMOS transistor whose gate is split into two parts  
20 each having a width of  $w/2$ ;  
said CMOS transistor whose source is split into two parts each having a width of  $w/2$ ;  
said two gate parts coupled together and to transistor gate output;  
25 said two source parts coupled together and to transistor source output.

5. A CMOS analog multiplexer circuit comprising:  
multiple series n-channel MOS transistor input  
switches;  
5 multiple series n-channel MOS transistor output  
switches;  
multiple n-channel MOS transistor pull-down switches;  
wherein  
said circuit is capable of multiplexing small level  
10 analog signals having amplitudes which are a small  
fraction of the power source;  
said circuit's parasitic capacitance is reduced by as  
much as 50%;  
said circuit's bandwidth is at least doubled; and  
15 said circuit prevents cross-signal feed-through from  
unselected inputs by shunting said feed-through to  
circuit ground.

6. The CMOS analog multiplexer circuit of Claim 5  
20 further comprising:  
the input of a first series input n-channel MOS  
transistor switch coupled to the first circuit input  
signal;  
the input of a second series input n-channel MOS  
25 transistor switch coupled to the second circuit  
input signal;

the input of a nth series input n-channel MOS transistor switch coupled to the nth circuit input signal;

5 the output of said first series input n-channel MOS transistor switch coupled to the input of a first output n-channel transistor switch and to the drain of a first n-channel MOS pull-down transistor;

10 the output of said second series input n-channel MOS transistor switch coupled to the input of a second output n-channel MOS transistor switch and to the drain of a second n-channel MOS pull-down transistor;

15 the output of said nth series input n-channel MOS transistor switch coupled to the input of a nth output n-channel MOS transistor switch and to the drain of a nth n-channel MOS pull-down transistor; the outputs of said first, second, and nth output n-channel MOS transistor switches coupled together and to the circuit output;

20 the sources of said first, second, and nth n-channel MOS pull-down transistors coupled to circuit ground; the gates of all said series input and output n-channel MOS transistor switches coupled to a logic control signal;

the gates of all said n-channel MOS pull-down transistors coupled to a logic control signal which is the complement of said series input/output transistor logic control signal.

5

7. A CMOS analog multiplexer circuit for use with differential input signals, comprising:

multiple positive and negative signal series n-channel MOS transistor input switches;

10 multiple positive and negative signal series n-channel MOS transistor output switches;

multiple positive and negative signal n-channel MOS transistor pull-down switches; wherein

said series positive and negative MOS transistor

15 switch's capacitance is reduced by at least 50% by means of eliminating the p-channel MOS transistors;

said circuit's bandwidth is at least doubled by means of eliminating said p-channel MOS transistors;

said circuit is capable of multiplexing small level

20 analog signals having amplitudes which are a small fraction of the power source; and

said circuit prevents cross-signal feed-through from unselected inputs by shunting said feed-through to circuit ground.

25

8. The CMOS analog multiplexer circuit of Claim 7  
further comprising:  
the positive inputs of multiple differential input  
signals coupled to corresponding multiple positive  
input series n-channel MOS transistor switches,  
5 respectively;  
the outputs of said multiple positive input series  
input n-channel MOS transistor switches coupled to  
the inputs of multiple positive output n-channel MOS  
transistor switches, respectively, and to the drains  
10 of multiple positive n-channel MOS pull-down  
transistors;  
the outputs of said multiple positive output  
n-channel MOS transistor switches coupled together  
and connected to the positive differential circuit  
15 output;  
the sources of said multiple positive n-channel MOS  
pull-down transistors coupled to circuit ground;  
the gates of all said positive series input and  
output n-channel MOS transistor switches coupled to  
20 a logic control signal;  
the gates of all said positive n-channel MOS pull-  
down transistors coupled to a logic control signal  
that is complementary to said positive series  
25 input/output logic control signal;

the negative inputs of multiple differential input signals coupled to corresponding multiple negative input series n-channel MOS transistor switches, respectively;

5 the outputs of said multiple negative input series input n-channel MOS transistor switches coupled to the inputs of multiple negative output n-channel MOS transistor switches, respectively, and to the drains of multiple negative n-channel MOS pull-down transistors;

10 the outputs of said multiple negative output n-channel MOS transistor switches coupled together and connected to the negative differential circuit output;

15 the sources of said multiple negative n-channel MOS pull-down transistors coupled to circuit ground; the gates of all said negative series input and output n-channel MOS transistor switches coupled to a logic control signal;

20 the gates of all said negative n-channel MOS pull-down transistors coupled to a logic control signal that is complementary to said negative series input/output logic control signal.

9. The CMOS analog multiplexer circuit of Claim 6 or 8 further comprising MOS transistor layouts having an even number of circuit fingers: wherein said MOS transistor's parasitic capacitance is further reduced by more than 50%; and said MOS transistor's bandwidth is further improved by at least two times.
10. The CMOS transistor layout of Claim 9 further comprising:
- a CMOS transistor whose architecture has the source and gate split in half such that it consists of the sequence: one-half source, one-half gate, drain, one-half gate, and one-half source;
  - said CMOS transistor whose drain capacitance is half that of a conventional CMOS transistor;
  - said CMOS transistor whose gate is split into two parts each having a width of  $w/2$ ;
  - said CMOS transistor whose source is split into two parts each having a width of  $w/2$ ;
  - said two gate parts coupled together and to transistor gate output;
  - said two source parts coupled together and to transistor source output.

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## ANALOG MULTIPLEXERS WITH CMOS CONTROL SIGNALS

### ABSTRACT

- 5 Analog multiplexer circuits with CMOS control signals and  
with low signal feed-through and high bandwidth are  
described. These circuits emphasize low parasitic  
capacitance through circuit layout techniques and the use  
of smaller size n-channel transistors where possible.
- 10 These circuits can be used for both single-ended and  
differential configurations. Two embodiments of the  
circuit are discussed allowing for optimal selection of  
multiplexers in application requirements ranging from  
lower-to-higher bandwidth and small-to-large input signal
- 15 size.

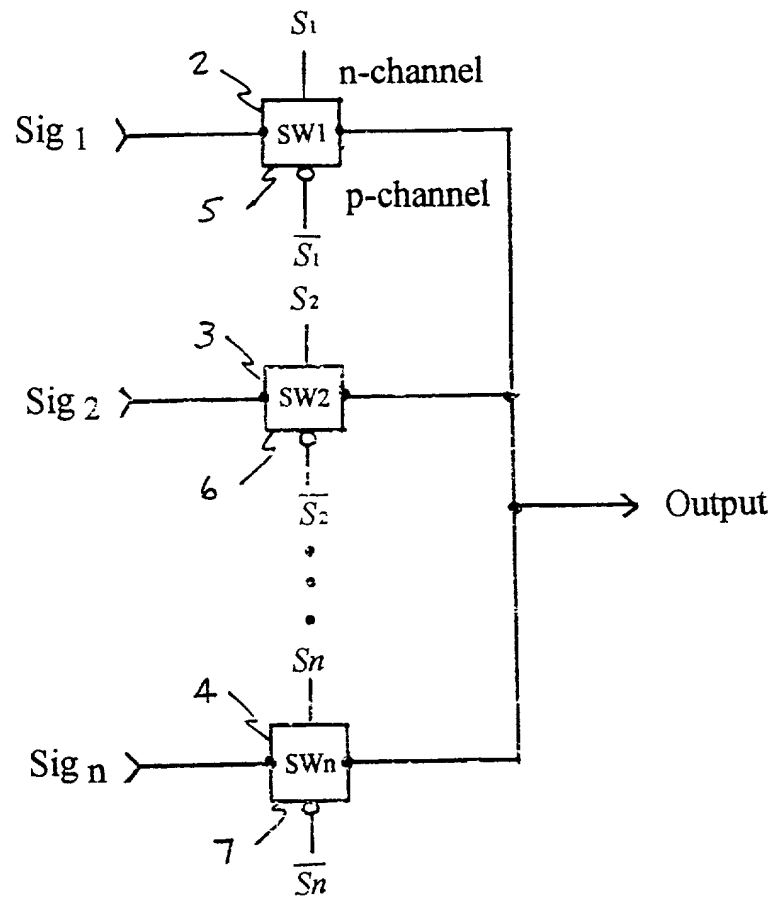


Fig. 1a

prior art

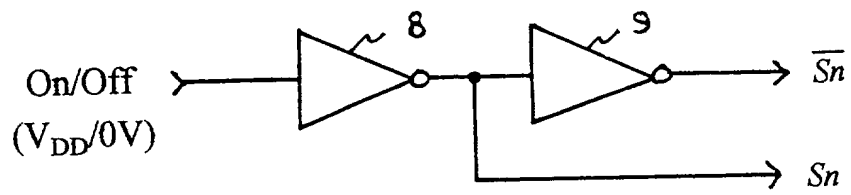


Fig. 1b

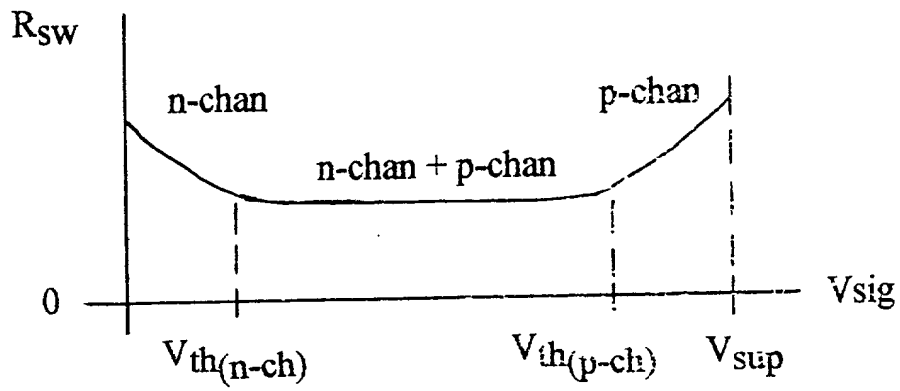


Fig. 1c

prior art

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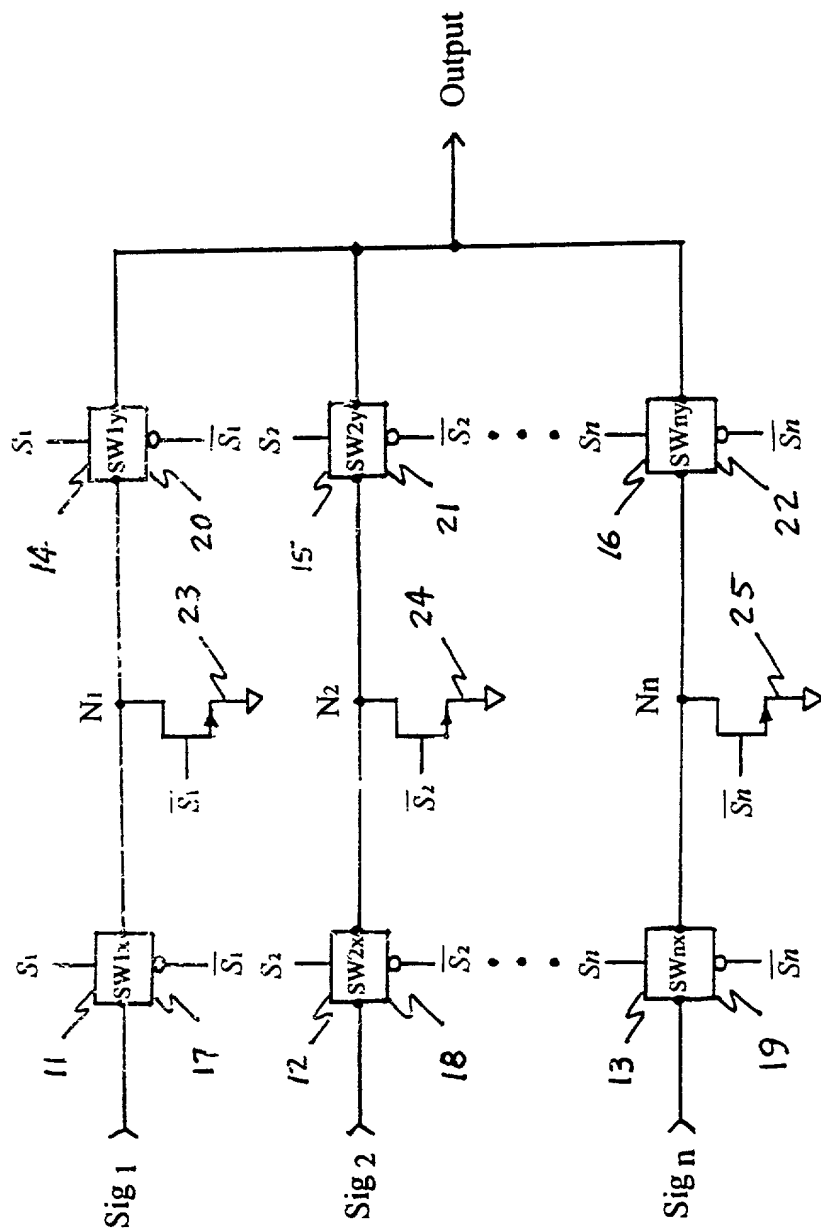


Fig. 2

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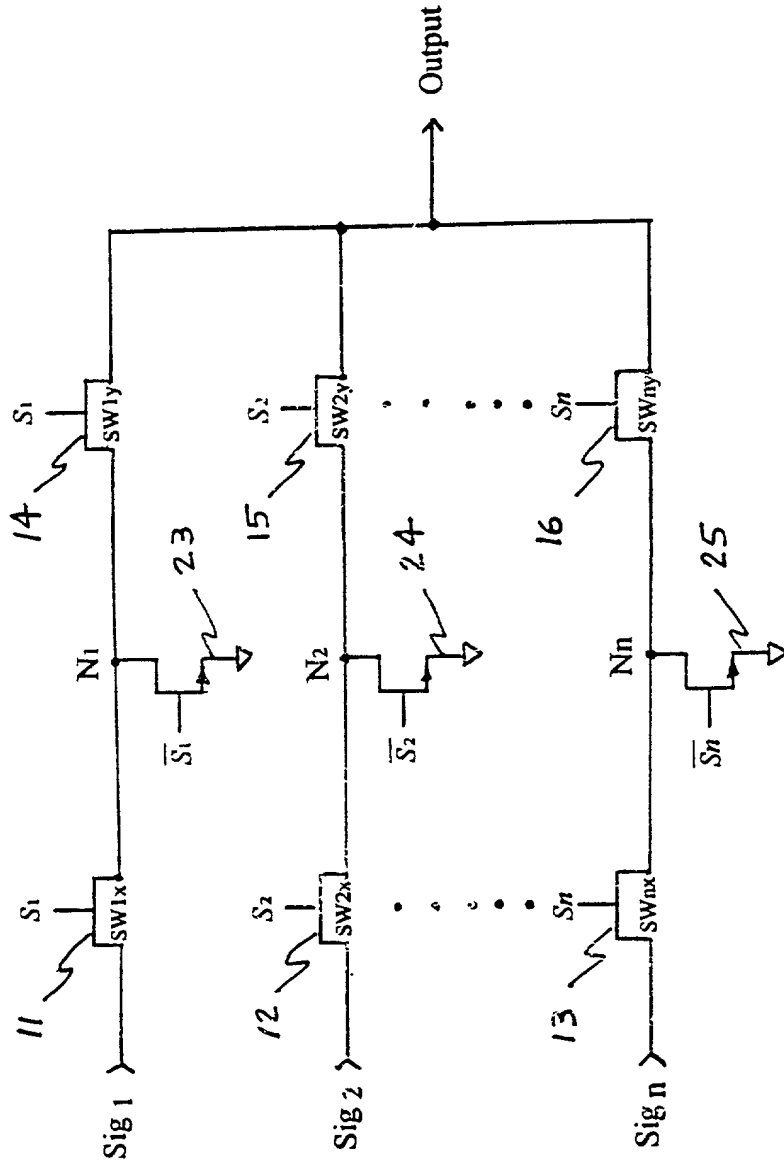


Fig. 3

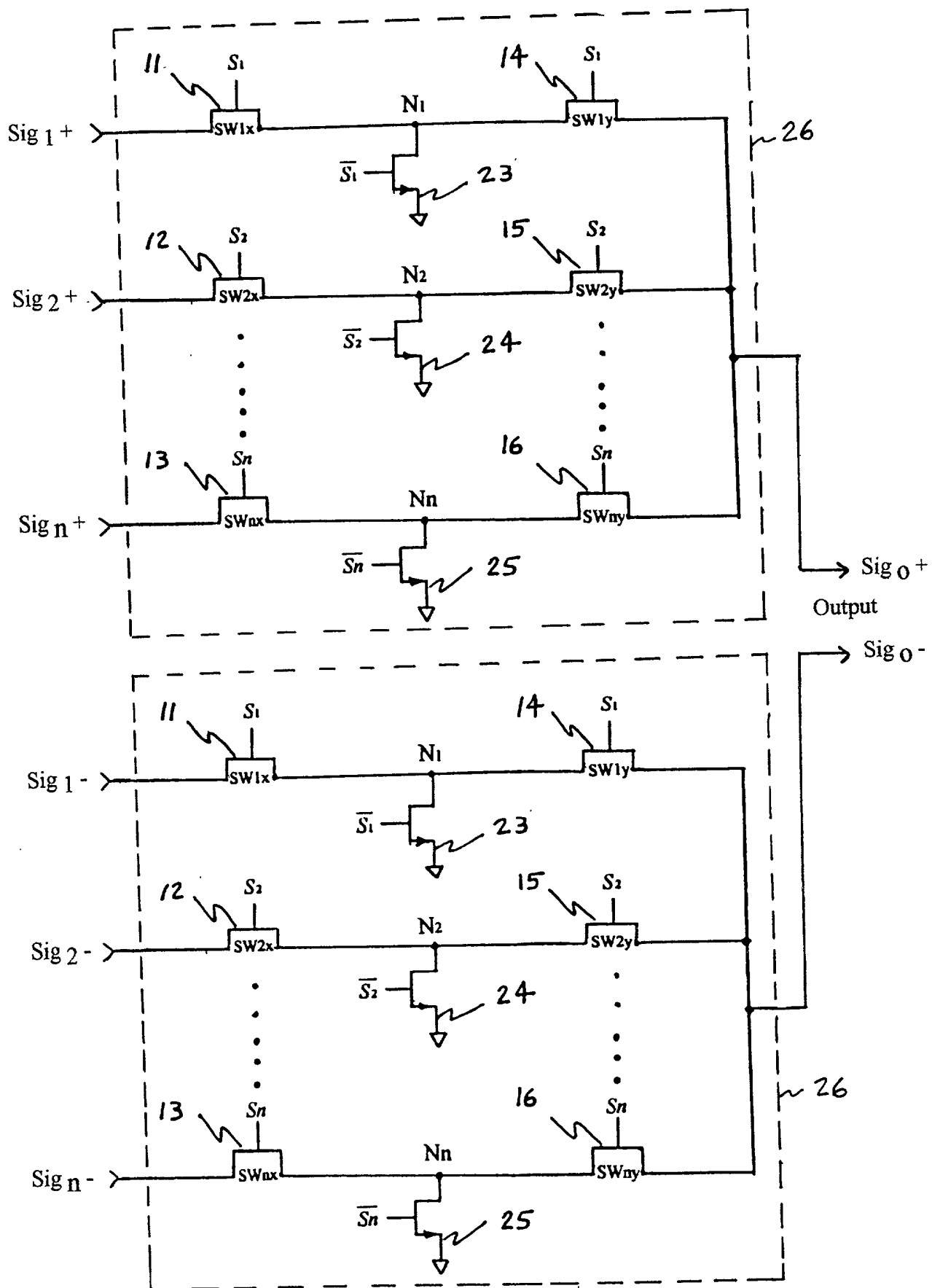


Fig. 4

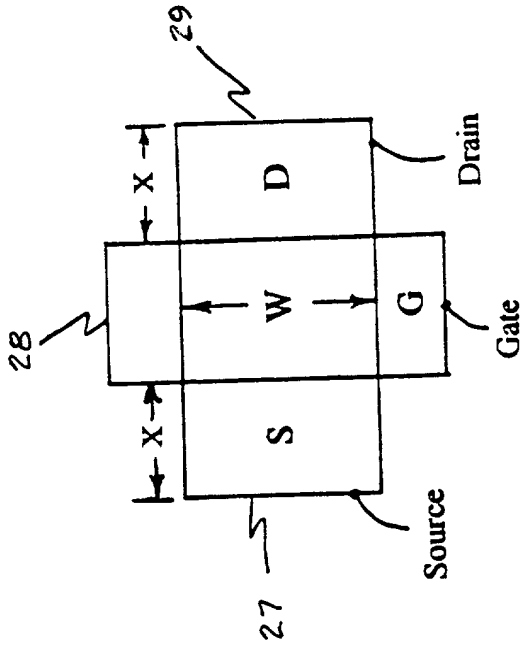


Fig. 5a

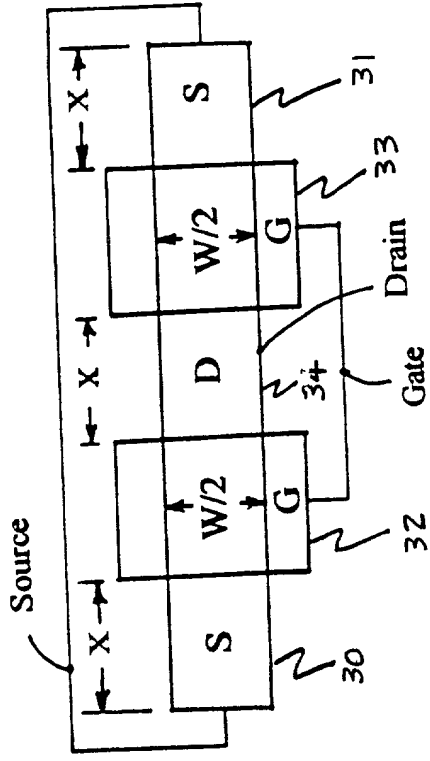
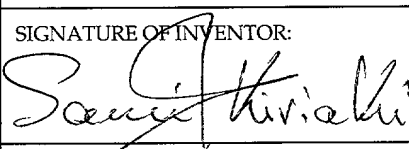


Fig. 5b

APPLICATION FOR UNITED STATES PATENT  
DECLARATION AND POWER OF ATTORNEY

As a below named inventor, I declare that my residence, post office address and citizenship are as stated below next to my name; that I verily believe that I am the original, first and sole inventor if only one name is listed below, or an original, first and joint inventor if plural inventors are named below, of the subject matter which is claimed and for which a patent is sought on the invention entitled as set forth below, which is described in the attached specification; that I have reviewed and understand the contents of the specification, including the claims, as amended by any amendment specifically referred to in the oath or declaration; that no application for patent or inventor's certificate on this invention has been filed by me or my legal representatives or assigns in any country foreign to the United States of America; and that I acknowledge my duty to disclose information which is material to the patentability of this application in accordance with Title 37, Code of Federal Regulations, section 1.56;

I further declare that all statements made herein of my own knowledge are true and that all statements made on information and belief are believed to be true, and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under section 1001 of Title 18 of the United States Code, and that such willful false statements may jeopardize the validity of the application or any patent issuing thereon.

TITLE OF INVENTION: <b>Analog Multiplexers With CMOS Control Signals</b>		
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SIGNATURE OF INVENTOR: 	SIGNATURE OF INVENTOR:	SIGNATURE OF INVENTOR:
DATE: 11/28/99	DATE:	DATE: